

EEE 324 Digital Signal Processing

# Lecture 4

Down-sampling

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#### Contents

• Down-sampling



• The sampling rate of a sequence can be reduced by defining a new sequence

$$x_d[n] = x[nM] = x_c(nMT)$$
(34)

Where M is a positive integer

• The system defined by Eq. (34) is called a *compressor* and is shown below.



•  $x_d[n]$  is identical to the sequence that would be obtained from  $x_c(t)$  by sampling with period T' = MT



• Moreover, if  $X_c(j\Omega) = 0$  for  $|\Omega| \ge \Omega_N$ , i.e.  $x_c(t)$  is a BL signal,

• then  $x_d[n]$  is an exact representation of  $x_c(t)$  if  $\pi/T' = \pi/MT \ge \Omega_N$ .

- The sampling rate can be reduced by a factor M without aliasing using two techniques: i.e.,
  - 1. if: the original sampling rate was at least M times the Nyquist rate (i.e.,  $\Omega_s \ge M 2 \Omega_0$ ), or
  - 2. if the bandwidth of the sequence is first reduced by a factor of M by DT filtering.



**Frequency Domain Relationship Between I/P and O/P of Compressor** 

- Since the compressor is a DT system, we will employ DTFT.
- The DTFT of  $x[n] = x_c(nT)$  is:

$$X(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c \left( j \left( \frac{\omega}{T} - \frac{2\pi k}{T} \right) \right)$$
(Recall Eq. (10))

The DTFT of 
$$x_d[n] = x[nM] = x_c(nT')$$
 with  $T' = MT$  is:  

$$\begin{bmatrix} X_d(e^{j\omega}) = \frac{1}{T'} \sum_{r=-\infty}^{\infty} X_c\left(j\left(\frac{\omega}{T'} - \frac{2\pi r}{T'}\right)\right) \end{bmatrix} \quad (35) \quad (\text{Recall Eq. (10)})$$

Putting the value of T',

$$\left|X_{d}\left(e^{j\omega}\right) = \frac{1}{MT} \sum_{r=-\infty}^{\infty} X_{c}\left(j\left(\frac{\omega}{MT} - \frac{2\pi r}{MT}\right)\right)\right| \quad (36)$$



**Frequency Domain Relationship Between I/P and O/P of Compressor** 

$$r = i + kM$$
,  $-\infty < k < \infty$ ,  $0 \le i \le M - 1$ 

$$\left[X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} \left[\frac{1}{T} \sum_{k=-\infty}^{\infty} X_c\left(j\left(\frac{\omega}{MT} - \frac{2\pi k}{T} - \frac{2\pi i}{MT}\right)\right)\right]\right]$$
(3)





**Frequency Domain Relationship Between I/P and O/P of Compressor** 

• The term inside the square brackets of Eq. (37) resemble Eq. (10) and can be written as:

$$X(e^{j(\omega-2\pi i)/M}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c \left( j \left( \frac{(\omega-2\pi i)}{MT} - \frac{2\pi k}{T} \right) \right)$$

*Hence*, *Eq*. (37)*can be expressed as*:

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X(e^{j(\omega-2\pi i)/M})$$
(38)



#### **Frequency Domain Relationship Between I/P and O/P of Compressor**

From Eq. (36), we can say that, in frequency domain, down-sampling by a factor of M can be thought of as being composed of:

- An *infinite* set of *copies* of  $X_c(j\Omega)$ , frequency *scaled* through  $\omega = \Omega T'$  and,
- *Shifted* by integer multiples of  $\frac{2\pi}{\tau'}$



**Frequency Domain Relationship Between I/P and O/P of Compressor** 

From Eq. (38), we can say that:

- in frequency domain, down-sampling by a factor of M produces M aliased copies of the DTFT  $X(e^{j\omega})$ .
- These copies are produced by:
  - *Stretching* the frequency axis by a factor of M,
  - and then *shifting* by  $2\pi i$ .



**Frequency Domain Relationship Between I/P and O/P of Compressor** Either interpretation makes it clear that:

•  $X_d(e^{j\omega})$  is periodic with period  $2\pi$ , and



Frequency Domain Relationship Between I/P and O/P of Compressor <u>Two ways of avoiding aliasing during downsampling</u>

1. Aliasing can be avoided by ensuring that  $X(e^{j\omega})$  is band-limited i.e.,  $X(e^{j\omega}) = 0, \qquad \omega_N \le |\omega| \le \pi$ 

And

$$\frac{2\pi}{M} \ge 2\omega_N$$

Or





**Frequency Domain Relationship Between I/P and O/P of Compressor** 





**Frequency Domain Relationship Between I/P and O/P of Compressor** 





#### **Frequency Domain Relationship Between I/P and O/P of Compressor**

## <u>Two ways of avoiding aliasing during downsampling</u>

2. Reduce the bandwidth of the signal x[n] first by using an LP filter with cut-off frequency  $\pi/M$  and then down-sample.

For the example shown on the right, the following parameters are used:

$$\Omega_s = 4\Omega_N, M = 2$$
$$\Omega_N = \frac{1}{4}\Omega_s$$
$$\omega_N = \frac{\pi}{2}$$



Figure 4.22 (a)-(c) Downsampling with aliasing. (d)-(f) Downsampling with prefiltering to avoid aliasing.





**Frequency Domain Relationship Between I/P and O/P of Compressor** 

• If the original signal was band-limited, such that:

$$X(e^{j\omega}) = 0, \quad \frac{\pi}{M} \le |\omega| \le \pi$$

there would be no aliasing.

- To avoid aliasing, we need to first low-pass filter, and then down-sample.  $H_d(e^{j\omega}) = \begin{cases} 1, & |\omega| < \pi/M \\ 0, & \frac{\pi}{M} \le |\omega| \le \pi \end{cases}$
- In practice, due to unfavourable properties of  $h_d[n]$  (infinite impulse response, slow decay), we do not apply this idealized filter.



#### **Frequency Domain Relationship Between I/P and O/P of Compressor**

• A general system for down-sampling by a factor of M is shown below.



- Such a system is called a *decimator*, and
- down-sampling by low-pass filtering followed by compression is called *decimation*.



### Reading

#### Section 4.6.1 (Oppenheim)

#### **Practice Problems**

**Problems: 4.14 – 4.18, 4.26, 4.27 (Oppenheim)** 

